Out-of-plane precession of an in-plane magnetized free layer submitted to the spin-transfer torque of a perpendicular polarizer: An analytical perturbative approach

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An analytical perturbative approach is presented to calculate the out-of-plane precessional motion of the magnetization of an in-plane free layer submitted to the spin torque of a perpendicular polarizer in the macrospin model. We consider the effects of the uniaxial anisotropy field, of an in-plane applied field, and of the perturbation due to the in-plane reference layer on the otherwise circular and monochromatic oscillations of the free layer. We calculate the frequency change due to these perturbations, the amplitude of the second harmonics, and the critical current for the existence of oscillations. This approach is rather general in the treatment of harmonics in spin torque oscillators.

I. INTRODUCTION

Spin torque oscillators (STOs) are spintronics nanodevices that produce radio frequencies in the gigahertz range with a dc current.[1–3] Due to spin-transfer torque, a spin polarized current flowing through a free magnetized layer is able to modify its magnetization orientation, producing self-sustained oscillations of the free-layer magnetization[4] that can be measured through a variation of the giant magnetoresistance (GMR), for spin valves, or the tunnel magnetoresistance (TMR) for magnetic tunnel junctions (MTJ). While most devices take advantage of in-plane precession (IPP) of the free-layer magnetization,[5] recent studies have shown the possibility of fabricating STOs in which the free-layer magnetization undergoes out-of-plane precessional motion (OPP).[6,7] Such devices are based on an in-plane magnetized MTJ, comprised of an in-plane magnetized free layer and an in-plane analyzer producing a magnetoresistive signal, with the addition of a perpendicular polarizer. The effect of the spin-transfer torque (STT) due to the perpendicular polarizer is to pull the free-layer magnetization out of plane, and due to the strong demagnetizing field, the free-layer magnetization then precesses around the direction normal to the plane. One advantage of the OPP over IPP is the large amplitude of the OPP oscillations, which gives a larger resistance variation, and therefore a larger RF output signal.

The most simple macrospin models of an in-plane MTJ with a perpendicular polarizer take into account only the demagnetizing field and the spin torque from the perpendicular polarizer,[8,9] as more general models do not seem to be solvable analytically. Here, we present a perturbative approach that allows to estimate the effect of the applied field, the anisotropy field, and the spin torque from the analyzer on the OPP state. We were able to compute the change in frequency due to these perturbations and the critical current for the existence of the OPP steady state submitted to these perturbations.

This approach is different from earlier studies of the STO dynamics[10–12] for which the dynamical steady state was calculated as a small deviation from the equilibrium state. It differs also from the planar approximation,[13] which applies in the assumption that the dynamics is almost in-plane. Here, the dynamical steady state is a deviation from another limit cycle with a simpler expression.

Summary of results. Let $z_0 < 1$ be the average out-of-plane angle of the free-layer magnetization, which is proportional to the applied current I: $z_0 = \frac{\hbar}{2e} \frac{\eta_{\mathrm{PERP}} I}{\alpha \mu_0 M_s^2 V}$. The main results of this paper are summarized in the table below, where we present the frequency change from the unperturbed frequency f_0, the amplitude of the second harmonic, and the critical current I_{c4}, the lower boundary for the existence of OPP, with respect to the perturbation due to an in-plane applied field, H_x or H_y, due to the in-plane uniaxial anisotropy field H_k, and due to the STT of the reference layer p_x:

	H_x, H_y	H_k	p_x
$(f - f_0)/f_0$	$-\frac{1}{2z_0^4}\left(\frac{H_{x/y}}{M_S}\right)^2$	$-\frac{1}{32z_0^4}\left(\frac{H_k}{M_S}\right)^2$	$\frac{1}{2}\left(\frac{p_x}{M_S}\right)^2$
2nd harmonic	$H_{x/y}/z_0^2$	0	p_x/z_0
$I_{c4} \sim \cdots$	$\alpha\sqrt{H_{x/y}M_s}$	$\alpha\sqrt{H_k M_s}$	0

II. MACROSPIN EQUATION AND SOLVABLE CASE

A. LLGS equation

In the system described in Fig. 1(a), the free-layer magnetization $\boldsymbol{m}$ is the solution of the following Landau-Lifschitz-Gilbert-Slonczewski (LLGS) equation in the macrospin approximation:
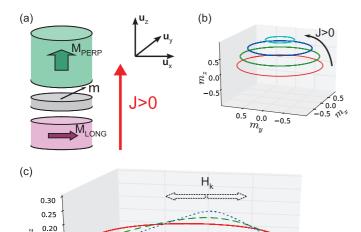
$$\frac{d\boldsymbol{m}}{dt} = -\mu_0\gamma(\boldsymbol{m} \times \boldsymbol{H}_{\mathrm{eff}}) + \alpha\left(\boldsymbol{m} \times \frac{d\boldsymbol{m}}{dt}\right) - \mu_0\gamma[\boldsymbol{m} \times (\boldsymbol{m} \times \boldsymbol{M}_{\mathrm{pol}})]. \quad (1)$$

Here, γ is the gyromagnetic ratio, μ_0 is the vacuum permeability, and α is the Gilbert damping constant. $\frac{d\boldsymbol{m}}{dt}$ is the time derivative of the free-layer magnetization $\boldsymbol{m}$. $\boldsymbol{H}_{\mathrm{eff}}$ is the effective field that derives from the magnetic free energy E_0, and $\boldsymbol{M}_{\mathrm{pol}}$ is the spin torque vector:

$$\boldsymbol{H}_{\mathrm{eff}} = -\frac{1}{\mu_0 M_S V}\frac{\partial E_0}{\partial \boldsymbol{m}},$$

$$\boldsymbol{M}_{\mathrm{pol}} = -a_{J\ \mathrm{LONG}}\boldsymbol{m}_{\mathrm{LONG}} + a_{J\ \mathrm{PERP}}\boldsymbol{m}_{\mathrm{PERP}}.$$

M_S is the saturation magnetization and V is the volume of the layer. $\boldsymbol{m}_{\mathrm{LONG}}$ (respectively, $\boldsymbol{m}_{\mathrm{PERP}}$) is the unit vector

FIG. 1. (Color online) (a) Schematics of the in-plane MTJ with a perpendicular polarizer. The applied field can be in any direction. (b) Unperturbed trajectory for different current density J: 2, 6, 10, and 14×10^{11} A/m^2. (c) Perturbed trajectory for $J = 1.5 \times 10^{11}$ A/m^2 and different anisotropy field H_k: 0, 24, and 48 kA/m.

collinear to the magnetization of the longitudinal (respectively, perpendicular) polarizer. Namely, $\boldsymbol{m}_{\mathrm{LONG}} = \boldsymbol{u}_x$ and $\boldsymbol{m}_{\mathrm{PERP}} = \boldsymbol{u}_z$. There is a minus sign in front of the term coming from the longitudinal polarizer because the perpendicular polarizer is situated above the free layer, whereas the perpendicular polarizer is below. For $i = (\mathrm{LONG}, \mathrm{PERP})$, a_{Ji} is the spin torque amplitude due to the corresponding polarizing layer. Its expression is given by[14]

$$a_{Ji} = \frac{\hbar}{2e} \frac{J}{\mu_0 M_S t} \frac{\eta_i}{1 + \lambda_i \boldsymbol{m} \cdot \boldsymbol{m}_i},$$

where t is the free-layer thickness, η_{LONG} and η_{PERP} are the spin polarizations of the polarizing layers, and we call λ_{LONG} and λ_{PERP} the spin polarization asymmetries. The spin polarization asymmetry depends on the nature of the nonmagnetic spacer layer separating the free layer from the polarizer, it is almost zero for a tunnel barrier and around 0.5 for a copper spacer.[14,15]

Equation (1) can be written in a more convenient way, with $\gamma_0 = \mu_0 \gamma / (1 + \alpha^2)$, as

$$\frac{d\boldsymbol{m}}{dt} = -\gamma_0 \boldsymbol{m} \times \boldsymbol{H}_{\mathrm{eff}} - \gamma_0 \alpha \boldsymbol{m} \times (\boldsymbol{m} \times \boldsymbol{H}_{\mathrm{eff}})$$
$$+ \gamma_0 \alpha \boldsymbol{m} \times \boldsymbol{M}_{\mathrm{pol}} - \gamma_0 \boldsymbol{m} \times (\boldsymbol{m} \times \boldsymbol{M}_{\mathrm{pol}}). \quad (2)$$

For convenience, we introduce P_0, which we call the spin torque potential, given by

$$P_0 = -\frac{\hbar}{2e} J S \eta_{\mathrm{LONG}} \frac{\ln(1 + \lambda_{\mathrm{LONG}} \boldsymbol{m}_{\mathrm{LONG}} \cdot \boldsymbol{m})}{\lambda_{\mathrm{LONG}}}$$
$$+ \frac{\hbar}{2e} J S \eta_{\mathrm{PERP}} \frac{\ln(1 + \lambda_{\mathrm{PERP}} \boldsymbol{m}_{\mathrm{PERP}} \cdot \boldsymbol{m})}{\lambda_{\mathrm{PERP}}}.$$

Here, S is the free-layer area through which the current flows. The spin torque vector $\boldsymbol{M}_{\mathrm{pol}}$ derives from P_0:

$$\boldsymbol{M}_{\mathrm{pol}} = \frac{1}{\mu_0 M_S V} \frac{\partial P_0}{\partial \boldsymbol{m}}.$$

P_0 has the advantage of being coordinate independent, making the change of variables easier. For $\lambda_{\mathrm{LONG}} = 0$, P_0 is defined by taking the limit when $\lambda_{\mathrm{LONG}} \to 0$ and similarly for $\lambda_{\mathrm{PERP}} = 0$.

We also introduce the two scalar functions $\mathcal{H}$ and $\mathcal{G}$, which are base invariant:

$$\mathcal{H} = \gamma_0 \frac{E_0 + \alpha P_0}{\mu_0 M_S V}, \quad \mathcal{G} = \gamma_0 \frac{\alpha E_0 - P_0}{\mu_0 M_S V}. \quad (3)$$

Equation (2) becomes

$$\frac{d\boldsymbol{m}}{dt} = \boldsymbol{m} \times \nabla_{\boldsymbol{m}} \mathcal{H} + \boldsymbol{m} \times (\boldsymbol{m} \times \nabla_{\boldsymbol{m}} \mathcal{G}). \quad (4)$$

$\mathcal{H}$ is equivalent to a Hamiltonian, it is a conservative term, whereas $\mathcal{G}$ is a dissipative term. Without spin torque, $\mathcal{H}$ and $\mathcal{G}$ are proportional, their ratio being α, but this simple relation is not valid anymore when we take into account the spin-transfer torque.

Using the condensed notation $F_X = \frac{\partial F}{\partial X}$, Eq. (4) simplifies in spherical coordinates:

$$m_x = \sin\theta \cos\phi,$$
$$m_y = \sin\theta \sin\phi,$$
$$m_z = \cos\theta,$$
$$\dot{\theta} = -\frac{\mathcal{H}_\phi}{\sin\theta} - \mathcal{G}_\theta, \quad \dot{\phi}\sin\theta = \mathcal{H}_\theta - \frac{\mathcal{G}_\phi}{\sin\theta}. \quad (5)$$

In the following, we consider a thin free layer. Its magnetic free energy is the sum of three contributions, respectively, the demagnetizing energy, a uniaxial anisotropy energy, and the Zeeman energy:

$$E_0 = \tfrac{1}{2}\mu_0 V M_S^2 m_z^2 + K_u V(1 - m_x^2) - \mu_0 M_S V \boldsymbol{m} \cdot \boldsymbol{H}_a.$$

K_u is a uniaxial anisotropy energy constant, related to the uniaxial anisotropy effective field $H_k = \frac{2K_u}{\mu_0 M_S}$. The uniaxial anisotropy accounts for magnetocrystalline and shape anisotropy and it favors x-axis orientation, which is the easy axis of magnetization. Here, the applied field $\boldsymbol{H}_a = (H_x, H_y, H_z)$ is in an arbitrary direction, with norm H_a.

The two scalar functions defined above, the conservative $\mathcal{H}$ and dissipative $\mathcal{G}$ terms, have the following expressions in spherical coordinates:

$$\mathcal{H} = \gamma_0 \left[\frac{M_S}{2}\cos^2\theta - \frac{H_k}{2}\sin^2\theta \cos^2\phi - H_x \sin\theta \cos\phi \right.$$
$$- H_y \sin\theta \sin\phi - H_z \cos\theta$$
$$\left. + \frac{\alpha p_z}{\lambda_z}\ln(1 + \lambda_z \cos\theta) - \frac{\alpha p_x}{\lambda_x}\ln(1 + \lambda_x \sin\theta \cos\phi) \right],$$

$$\mathcal{G} = \gamma_0 \left[\frac{p_x}{\lambda_x} \ln(1 + \lambda_x \sin\theta \cos\phi) - \frac{p_z}{\lambda_z} \ln(1 + \lambda_z \cos\theta) \right.$$
$$+ \alpha \frac{M_S}{2} \cos^2\theta - \alpha \frac{H_k}{2} \sin^2\theta \cos^2\phi$$
$$\left. - \alpha H_x \sin\theta \cos\phi - \alpha H_y \sin\theta \sin\phi - \alpha H_z \cos\theta \right].$$

Here, $p_x = \frac{\hbar}{2e} \frac{J\eta_{\text{LONG}}}{\mu_0 M_S t}$, $p_z = \frac{\hbar}{2e} \frac{J\eta_{\text{PERP}}}{\mu_0 M_S t}$, $\lambda_x = \lambda_{\text{LONG}}$, and $\lambda_z = \lambda_{\text{PERP}}$ for compactness. The demagnetizing energy is assumed, henceforth, to be the dominant term of the magnetic free energy. In fact, we suppose that $M_S \gg H_k, H_a, a_J$, which is generally the case for thin films in nanopillars. Therefore the dominant term of the conservative part $\mathcal{H}$ comes from the demagnetizing energy $\frac{M_S}{2} \cos^2\theta$. The dissipative term $\mathcal{G}$ is smaller compared to the conservative part, by a factor α, and it is dominated by the spin torque $p_z \cos\theta/(1 + \lambda_z \cos\theta)$ and the damping of the demagnetizing field $\alpha \frac{M_S}{2} \cos^2\theta$. The uniaxial anisotropy field and the Zeeman field give a smaller contribution. As shown in Fig. 1(c), changing the anisotropy field, for instance, does not change much the magnetization trajectory. In fact, the trajectory of the dynamical system with a uniaxial anisotropy, an applied field, or an analyzer spin torque is very close to the trajectory with none of them. Hence it is reasonable to study first the simplest dynamical system describing a limit cycle, and then to treat the addition of other terms as perturbations around this limit cycle.

B. Unperturbed system

In a first attempt to describe the out-of-plane precession state of the free-layer magnetization with a perpendicular polarizer, we will consider only the demagnetization energy $\frac{M_S}{2} \cos^2\theta$ in the total energy, and only the z component of the spin torque p_z, in the particular case of an MTJ ($\lambda_z = 0$ and $P_0 = p_z m_z$).[8,16] Neglecting other terms, the trajectory is exactly circular [see Fig. 1(c)], and because they are the dominant terms, it is a good approximation of the complicated trajectory including all the terms. In this framework, the Hamiltonian and dissipative terms have simple expressions:

$$\mathcal{H} = \frac{\gamma_0 M_S}{2} \cos^2\theta + \alpha \gamma_0 p_z \cos\theta,$$
$$\mathcal{G} = \alpha \frac{\gamma_0 M_S}{2} \cos^2\theta - \gamma_0 p_z \cos\theta,$$

The equation of motion then writes as:

$$\dot\theta = \alpha\gamma_0 M_S \cos\theta \sin\theta - \gamma_0 p_z \sin\theta,$$
$$\dot\phi \sin\theta = -\gamma_0 M_S \cos\theta \sin\theta - \alpha\gamma_0 p_z \sin\theta. \tag{6}$$

We set: $\omega_0 = \frac{\gamma_0 p_z}{\alpha}$ and $z_0 = \frac{p_z}{\alpha M_S}$.
With the change of variables $z = \cos\theta$, Eq. (6) becomes

$$\dot z = -\alpha\omega_0(1 - z^2)\left(\frac{z}{z_0} - 1\right),$$
$$\dot\phi = -\omega_0\left(\frac{z}{z_0} + \alpha^2\right). \tag{7}$$

This equation can be solved explicitly for the variable z by separation of variables and using partial fraction

decomposition:

$$\left| \frac{z - z_0}{z_i - z_0} \right| \left| \frac{(1 + z)(1 - z_i)}{(1 - z)(1 + z_i)} \right|^{z_0/2} \sqrt{\frac{1 - z_i^2}{1 - z^2}} = e^{-\beta_0\omega_0 t}.$$

Here, z_i and ϕ_i are the initial out-of-plane magnetization and in-plane phase. The initial time is taken to be zero. We set $\beta_0 = \alpha(1 - z_0^2)/z_0$, with the inverse relaxation time $\beta_0\omega_0 = \alpha M_S\gamma_0(1 - z_0^2)$ always positive. As the left-hand side tends towards zero when time tends to infinity, the right-hand side must also tend towards zero. Except for the degenerate case of $z_i = \pm 1$, this condition means that $z \to z_0$ for long-term time. Therefore after a certain time, the condition $|z - z_0| \ll 1$ will be fulfilled, and we can derive an approximation for the out-of-plane magnetization z and the phase ϕ:

$$z - z_0 = (z_i - z_0)e^{-\beta_0\omega_0 t},$$
$$\phi - \phi_i = -\omega_0(1 + \alpha^2)t + \frac{z_i - z_0}{z_0\beta_0}(e^{-\beta_0\omega_0 t} - 1). \tag{8}$$

Equation (8) describes the unperturbed out-of-plane precession of the magnetization. In permanent regime, the precession is characterized by a constant out-of-plane magnetization $m_z = z_0$ and a constant frequency f_0, which is proportional to the applied current J[9] [see Fig. 1(b)]:

$$f_0 = \frac{\omega_0(1 + \alpha^2)}{2\pi} = \frac{\gamma\hbar\eta_{\text{PERP}}}{4\pi e\alpha M_S t} J. \tag{9}$$

This trajectory is in fact an attractive limit-cycle of the dynamical system described by Eq. (7). In the general case, the study of limit-cycle stability requires to define a Poincaré map but in the present case, where the geometrical space is two dimensional, one can directly extract the eigenvalue of the linearized Poincaré map, also known as multiplier μ_1:

$$\mu_1 = \exp\left\{ \int_0^T \text{div } \boldsymbol{F}[\boldsymbol{x}_0(t)]dt \right\}, \tag{10}$$

where T is the period of the limit cycle, $\boldsymbol{F}(x)$ is the vector field of the dynamical system $\dot{\boldsymbol{x}} = \boldsymbol{F}(\boldsymbol{x})$, and $\boldsymbol{x}_0(t)$ is a parameterization of the limit cycle. Expressed in spherical coordinates, Eq. (10) can be simplified with the expression of the vector field taken from the left-hand side of Eq. (5):

$$\mu_1 = \exp\left\{ -\int_0^T \left[\cot\theta \, \mathcal{G}_\theta(\boldsymbol{x}_0(t)) \right.\right.$$
$$\left.\left. + \mathcal{G}_{\theta\theta}(\boldsymbol{x}_0(t)) + \frac{\mathcal{G}_{\phi\phi}(\boldsymbol{x}_0(t))}{\sin^2\theta} \right] dt \right\}. \tag{11}$$

Notice that only the dissipative term appears in this expression. In the simplified case of Eq. (7), the time-parameterized spherical equation of the limit cycle is $\boldsymbol{x}_0(t) = [\theta_0, -\omega_0(1 + \alpha^2)t]$ ($z_0 = \cos\theta_0$). The expression of div $\boldsymbol{F}$ is then $-\gamma_0 p_z \frac{1 + 2z_0\cos\theta - 3\cos^2\theta}{z_0}$ and the period $T = 1/f_0 = \frac{2\pi}{\omega_0(1+\alpha^2)}$. The multiplier of the Poincaré map is then $\mu_1 = \exp(-\frac{2\pi\beta_0}{1+\alpha^2})$. The multiplier is the relaxation rate towards the limit cycle from a state slightly deviated, for instance, by thermal fluctuations.[16] Because μ_1 is less than unity in absolute value, the limit cycle is hyperbolically stable.

C. High-symmetry system

In addition to the previous simple case, we can add some terms that present the same symmetry as the unperturbed equation, as treated in previous papers.[8,16] We consider terms that are independent of the in-plane angle ϕ: the spin torque amplitude is angle dependent ($\lambda_z \neq 0$) and there is an out-of-plane applied field H_z. The out-of-plane applied can describe the stray field created by the perpendicular polarizer, or an external applied field. Eq. (5) is then

$$\dot{\theta} = \alpha \gamma_0 M_S \cos\theta \sin\theta - \frac{\gamma_0 p_z \sin\theta}{1 + \lambda_z \cos\theta} - \alpha \gamma_0 H_z \sin\theta,$$
(12)

$$\dot{\phi} \sin\theta = -\gamma_0 M_S \cos\theta \sin\theta - \frac{\alpha \gamma_0 p_z \sin\theta}{1 + \lambda_z \cos\theta} + \gamma_0 H_z \sin\theta.$$

As for the unperturbed system, the vector field is only θ-dependent, hence we can solve analytically the system. We expect to find at least one limit cycle, so we directly look for the permanent regime solution, with a constant angle θ. By changing the variables $z = \cos\theta$ and using the parameters ω_0, z_0, and $\nu_z = \gamma_0 H_z / \omega_0$, one can directly find the (nontrivial) solutions (z_+, z_-) of $\dot{z} = 0$:

$$z_\pm = -\frac{1 - \lambda_z z_0 \nu_z}{2\lambda_z} \pm \frac{1}{2\lambda_z} \sqrt{(1 + \lambda_z z_0 \nu_z)^2 + 4\lambda_z z_0}.$$

The positive solution z_+ tends toward z_0 when the parameters λ and ν_z tend toward zero: it is the perturbed solution. As we are interested here in solutions with a small out-of-plane magnetization z for application in precessional switching, we will focus only on the z_+ solution. Notice that z_- does not always exist (if $|z_-| > 1$ for instance), and when it does, it does not always describe an attractive limit cycle. With the constant out-of-plane magnetization z_+, the frequency of the precession is also changed compared to the unperturbed frequency $f_0 = \omega_0(1 + \alpha^2)/2\pi$:

$$f = \frac{f_0}{1 + \lambda_z z_+}.$$

For a small λ_z parameter, we obtain the approximation by Taylor expansion $f = f_0[1 - \lambda_z z_0(1 + \nu_z)]$. See Fig. 2 for comparison of the analytical expression of the frequency with $\lambda \neq 0$ and numerical simulations.

The high-symmetry system is not of real interest for applications in precessional switching, as an important parameter is omitted, the uniaxial anisotropy. However, since the approximation at order 1 is close to the exact solution, this gives credit to this approach and encourages the investigation of other additional perturbative terms.

The stability of the out-of-plane precession state (OPP) is insured by the condition $\mu_1 < 1$. It also means that if LLGS (6) is slightly modified, for instance, by adding the term associated with a uniaxial anisotropy, the OPP state will remain a limit cycle, for small enough perturbations.[16] In the following section, we will start from Eq. (6), henceforth called the unperturbed case, and study the general case with the general assumption that the solutions are very close to the unperturbed case. The unperturbed dynamical system will be modified by perturbations, namely, uniaxial anisotropy field, applied field, and longitudinal polarizer.
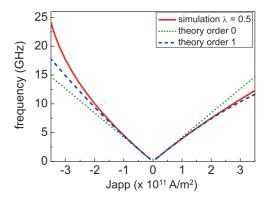

FIG. 2. (Color online) Oscillation frequency vs applied current density J. Comparison between macrospin simulation with $\lambda_z = 0.5$ (red full line) and analytical expressions at order 0 in λ_z (green dotted line) and order 1 (blue dashed line). The other simulation parameters are $\alpha = 0.02$, $M_S = 1.2 \times 10^6$ A/m, $\eta_{\text{PERP}} = 0.6$, $t = 3$ nm, and $H_z = 0$.

III. PERTURBATION THEORY

A. Uniaxial anisotropy

Compared to the previous case, including the uniaxial anisotropy term in Eq. (5) makes the vector field not only θ-dependent but also ϕ-dependent. Exact analytical solutions are then much more difficult to obtain, so we use a perturbative approach to approximate the solution,[17] using a Lindstedt's expansion.[18] When a periodic dynamical system is perturbed, the perturbated solution will also be periodic, by continuity. The perturbated solution is approximated by a generalized asymptotic expansion that preserves the periodicity of the solution, the Lindstedt's series for which the frequency is also expanded with respect to the perturbation parameter ϵ. The only assumption on the perturbation parameter is that it is small compared to the nonperturbed frequency, hence the equations are normalized to the frequency. Thus the approximation error is of order $O(\epsilon)$ on a time scale of $O(1/\epsilon)$, which is sufficient if the functions are periodic with a normalized period of 1.

On top of the uniaxial anisotropy perturbation, all the following cases will also be solved using this approach. From the unperturbed Eq. (6), we add the uniaxial anisotropy term:

$$\dot{\theta} = \gamma_0 \left[\alpha M_S \cos\theta \sin\theta - p_z \sin\theta - \frac{H_k}{2} \sin\theta \sin(2\phi) \right.$$
$$\left. + \alpha \frac{H_k}{2} \sin\theta \cos\theta (1 + \cos(2\phi)) \right],$$

$$\dot{\phi} \sin\theta = \gamma_0 \left[-M_S \cos\theta \sin\theta - \alpha \frac{H_k}{2} \sin\theta \sin(2\phi) \right.$$
$$\left. - \alpha p_z \sin\theta - \frac{H_k}{2} \sin\theta \cos\theta (1 + \cos(2\phi)) \right].$$ (13)

We use again the notation ω_0, z_0, and ϵ defined below. For clarity, the time variable is changed to τ:

$$\epsilon = \frac{\alpha}{1 + \alpha^2} \frac{H_k}{2 p_z}, \quad \tau = \omega_0(1 + \alpha^2) t.$$

The time derivative $\dot{X}$ is also changed to X', which means a derivative with respect to τ. We define $r_0 = \sqrt{1 - z_0^2}$ and $\beta = \alpha r_0^2/[z_0(1 + \alpha^2)]$. As $\alpha \ll 1$, the parameter β is very close to β_0 previously defined. For convenience, we use the functions

$$g_\theta(\theta) = \frac{\alpha}{1 + \alpha^2} \sin\theta \left(\frac{\cos\theta}{z_0} - 1\right),$$

$$h_\phi(\theta) = -\frac{\cos\theta + z_0\alpha^2}{z_0(1 + \alpha^2)} \sin\theta. \qquad (14)$$

We also define two perturbation functions, which in the case of the uniaxial anisotropy have the following form:

$$f_\theta^\epsilon(\theta,\phi) = -\sin\theta \sin(2\phi) + \alpha \sin\theta \cos\theta(1 + \cos(2\phi)),$$

$$f_\phi^\epsilon(\theta,\phi) = -\sin\theta \cos\theta(1 + \cos(2\phi)) - \alpha \sin\theta \sin(2\phi). \qquad (15)$$

Hence Eq. (13) takes the simplified form

$$\theta' = g_\theta(\theta) + \epsilon f_\theta^\epsilon(\theta,\phi),$$

$$\phi' \sin\theta = h_\phi(\theta) + \epsilon f_\phi^\epsilon(\theta,\phi). \qquad (16)$$

To solve this system, the following assumptions are made: (i) $\epsilon \ll 1$ (see Appendix) and (ii) the functions $\theta(\tau)$ and $\phi(\tau)$, solutions of Eq. (16), are written as a power series of the parameter ϵ. In the framework of the Lindstedt's series, we also consider that the period of the limit cycle can be written as a power series of ϵ, to get rid of the secular terms. In fact, we change the time variable $\tau' = \omega\tau$. Hence ω, θ, and ϕ will be written (up to the second order) as

$$\theta = \theta_0 + \epsilon\theta_1 + \epsilon^2\theta_2 + \dots,$$

$$\phi = \phi_0 + \epsilon\phi_1 + \epsilon^2\phi_2 + \dots, \qquad (17)$$

$$\omega = 1 + \epsilon f_1 + \epsilon^2 f_2 + \dots.$$

The development of the calculations is made in Appendix. For the uniaxial anisotropy, we find the trajectory equations up to first order:

$$\theta_0(\tau') = \arccos(z_0), \qquad \phi_0(\tau') = -\tau',$$

$$\begin{cases} \theta_1(\tau') = a_0 + a_2\cos(2\tau') + b_2\sin(2\tau'), \\ \phi_1(\tau') = c_2\cos(2\tau') + d_2\sin(2\tau'), \end{cases} \qquad (18)$$

with the coefficients defined by:

$$a_0 = (1 + \alpha^2)z_0^2/r_0,$$

$$a_2 = \frac{r_0}{4 + \beta^2}(\alpha\beta z_0 - 2),$$

$$b_2 = \frac{r_0}{4 + \beta^2}(2\alpha z_0 + \beta),$$

$$c_2 = -\frac{(1 + \alpha^2)\beta^2/2 + \alpha\beta z_0 + 2\alpha^2}{\alpha(4 + \beta^2)},$$

$$d_2 = -\frac{2\alpha z_0 + \beta}{\alpha(4 + \beta^2)}.$$

The frequency f of the periodic orbit is changed only at the second order in the parameter ϵ:

$$f_1 = 0, \quad f_2 \approx -\frac{1 + 4z_0^2}{8z_0^2 + 2\alpha^2}. \qquad (19)$$
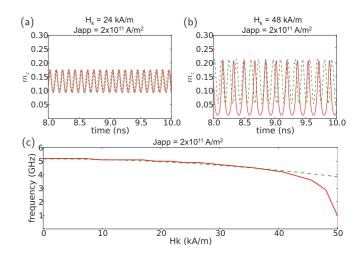

FIG. 3. (Color online) Macrospin simulation (red full line) and analytical expression at first order in ϵ (green dashed line). (a) and (b) Out-of plane component of the magnetization m_z vs time, with applied current density of $J_{app} = 2 \times 10^{11}$ A/m^2 ($z_0 = 0.13$) and with (a) $H_k = 24$ kA/m ($\epsilon = 0.05$) and (b) $H_k = 48$ kA/m ($\epsilon = 0.1$). (c) Frequency of the in-plane component m_x vs anisotropy field H_k. The other used parameters are given in the text.

With the expression of the unperturbed frequency f_0 found in Eq. (9), we obtain

$$f = f_0 \left(1 - \epsilon^2 \frac{1 + 4z_0^2}{8z_0^2 + 2\alpha^2}\right). \qquad (20)$$

These analytical expressions were compared to the results of 10-ns-long macrospin simulations. For the simulations, $\alpha = 0.02$, $M_S = 1.2 \times 10^6$ A/m, $t = 3$ nm, and $\eta_{PERP} = 0.2$.

Figures 3(a) and 3(b) show the out-of-plane component of the free-layer magnetization m_z in the permanent regime (8 ns after the application of the current pulse) for an applied current density of $J_{app} = 2 \times 10^{11}$ A/m^2. We compare the macrospin simulation in red full line with the analytical expression at first order in ϵ, in green dashed line. In Fig. 3(a), for $H_k = 24$ kA/m, corresponding to $\epsilon = 0.05$, the analytical expression is very similar to the macrospin simulations. The phase shift with the simulation is due to the transient regime that is not taken into account in the analytical expression. However, in Fig. 3(b), for $H_k = 48$ kA/m, and so $\epsilon = 0.1$, the trajectory from the macrospin simulation is not exactly sinusoidal and it cannot be described by the first-order approximation only. As seen also in Fig. 3(c), for larger value of the perturbation H_k, the frequency change is not well described by the second-order approximation anymore. Notice that the anisotropy induces a perturbation of twice the natural frequency because of the symmetry of the anisotropy term. The applied field and analyzer spin torque induce a perturbation at the natural frequency.

The trajectory of the magnetization is very close to the trajectory of the unperturbed system, as shown in Fig. 1(c). It validates the use of a perturbation expansion. The perturbed trajectory has a larger out-of-plane component m_z close to the easy-axis $\pm\boldsymbol{u}_x$ directions, and m_z is smaller close to the hard-axis directions $\pm\boldsymbol{u}_y$. The magnetization also spends more time along the $\boldsymbol{u}_x$ axis than along the $\boldsymbol{u}_y$ axis because the in-plane angle time derivative is smaller along the easy axis.

The magnetoresistive signal that can be measured with an in-plane analyzer depends on $m_x = \sin\theta \cos\phi$. Therefore the signal has a fundamental frequency and higher-order harmonics, of which the amplitude can be estimated as

$$m_x = \sin(\theta_0 + \epsilon\theta_1) \cos(-\tau + \epsilon\phi_1)$$
$$\approx r_0 \cos\tau + \epsilon(z_0\theta_1 \cos\tau + r_0\phi_1 \sin\tau).$$

Because θ_1 and ϕ_1 oscillate at twice the fundamental frequency, the second harmonic vanishes.

B. Applied field

We next discuss the perturbative effect of the applied field along the easy axis H_x and along the in-plane hard axis H_y. They are treated in the same section because they yield similar results. However, the respective perturbative developments were done separately.

We introduce the dimensionless variables

$$\nu_x = \frac{\alpha}{1+\alpha^2} \frac{H_x}{p_z}, \quad \nu_y = \frac{\alpha}{1+\alpha^2} \frac{H_y}{p_z}.$$

The perturbation functions on the angles θ and ϕ are, respectively,

$$f_\theta^{\nu_x}(\theta,\phi) = \alpha\cos\theta\cos\phi - \sin\phi,$$
$$f_\phi^{\nu_x}(\theta,\phi) = -\alpha\sin\phi - \cos\theta\cos\phi,$$
$$f_\theta^{\nu_y}(\theta,\phi) = \alpha\cos\theta\sin\phi + \cos\phi,$$
$$f_\phi^{\nu_y}(\theta,\phi) = \alpha\cos\phi - \cos\theta\sin\phi. \tag{21}$$

After solving the system up to second order, we find the expressions of the perturbed frequencies and the perturbed trajectories for the two directions of the applied field. In both cases,

$$f_1 = 0, \quad f_2 = -\frac{(1+\alpha^2)^2}{2z_0^2 + 2\alpha^2},$$
$$\begin{cases} \theta_1^i(\tau') = a_1^i \cos\tau' + b_1^i \sin\tau', \\ \phi_1^i(\tau') = c_1^i \cos\tau' + d_1^i \sin\tau', \end{cases} \tag{22}$$

with the coefficients for the field along the x axis given by:

$$a_1^x = -\frac{1 - \alpha\beta z_0}{1 + \beta^2},$$
$$b_1^x = \frac{\beta + \alpha z_0}{1 + \beta^2},$$
$$c_1^x = -\frac{\alpha\beta z_0 + \beta^2(1+\alpha^2) + \alpha^2}{\alpha r_0(1 + \beta^2)},$$
$$d_1^x = -\frac{\beta + \alpha z_0}{\alpha r_0(1 + \beta^2)},$$

and the coefficients for the field along the y axis given by:

$$a_1^y = \frac{\beta + \alpha z_0}{1 + \beta^2},$$
$$b_1^y = \frac{1 - \alpha\beta z_0}{1 + \beta^2},$$
$$c_1^y = -\frac{\beta + \alpha z_0}{\alpha r_0(1 + \beta^2)},$$
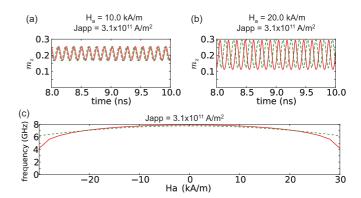$$d_1^y = \frac{\alpha\beta z_0 + \beta^2(1+\alpha^2) + \alpha^2}{\alpha r_0(1 + \beta^2)}.$$

FIG. 4. (Color online) Macrospin simulation (red full line) and analytical expression at first order in ν_x (green dashed line). (a) and (b) Out-of-plane component of the magnetization m_z vs time, with applied current density of $J_{\text{app}} = 3.1 \times 10^{11}$ A/m^2 ($z_0 = 0.19$) and with applied field (a) $H_x = 10$ kA/m ($\nu_x = 0.044$) and (b) $H_x = 20$ kA/m ($\nu_x = 0.088$). (c) Frequency of the in-plane component m_x vs applied field H_x.

Figure 4 shows the comparison with macrospin simulations using the same numerical values as previously, except that $J_{\text{app}} = 3.1 \times 10^{11}$ A/m^2 and with an in-plane applied field along $\boldsymbol{u}_x$. Figures 4(a) and 4(b) represent the out-of-plane component of the magnetization m_z after the transitory regime, for $H_x = 10$ kA/m ($\nu_x = 0.044$) and $H_x = 20$ kA/m ($\nu_x = 0.088$), respectively. In both cases, the analytical expression is in good agreement with the simulations. The frequency versus applied field, in Fig. 4(c), shows also a good agreement between the simulations and the approximation at second order in ν_x up to $H_x = 20$ kA/m.

Concerning the harmonics of the magnetoresistive signal, the second harmonic has a nonvanishing amplitude, using the formula of the previous section. Here, we consider that the applied current density is large enough so that $\beta \ll 1$ and small enough so that $z_0^2 \ll 1$. For $i = (x,y)$, the two directions of the in-plane applied field are

$$m_x = r_0 \cos\tau + \frac{\nu_i}{2}\left[\left(z_0 a_1^i - r_0 d_1^i\right)\cos(2\tau)\right.$$
$$\left. + \left(z_0 b_1^i + r_0 c_1^i\right)\sin(2\tau)\right].$$

Let ψ be a constant phase. The in-plane component m_x is approximately given by

$$m_x \approx r_0 \cos\tau + \frac{\nu_i}{2z_0}\cos(2\tau + \psi).$$

C. Spin torque from the reference layer

In this section, we focus on the perturbation due to the spin torque of the longitudinal polarizer, initially with $\lambda_x = 0$. We introduce the variable

$$\xi = \frac{\alpha}{1+\alpha^2} \frac{p_x}{p_z}.$$

The associated perturbation functions are given by

$$f_\theta^\xi(\theta,\phi) = -\cos\theta\cos\phi - \alpha\sin\phi,$$
$$f_\phi^\xi(\theta,\phi) = +\sin\phi - \alpha\cos\theta\cos\phi. \tag{23}$$

In experiments, the reference layer is most often separated from the free layer by a tunnel barrier, whereas the perpendicular polarizer is separated by a metallic barrier.[19,20] This choice is made in order to maximize the amplitude of the output TMR signal. Consequently, the ratio $r = p_x/p_z$ is greater than unity. The variable ξ can only be treated as a perturbation because of the α coefficient. Therefore the effect of the longitudinal polarizer on the OPP state is stronger than for other contributions. The solutions are given by

$$f_1 = 0, \qquad f_2 = \frac{(1+\alpha^2)^2 z_0^2}{2(1+\alpha^2 z_0^2)}, \qquad (24)$$

$$\begin{cases} \theta_1(\tau') = a_1 \cos\tau' + b_1 \sin\tau', \\ \phi_1(\tau') = c_1 \cos\tau' + d_1 \sin\tau', \end{cases} \qquad (25)$$

with the coefficients defined by:

$$a_1 = -\frac{\alpha + \beta z_0}{1+\beta^2},$$
$$b_1 = \frac{\alpha\beta - z_0}{1+\beta^2},$$
$$c_1 = \frac{\alpha + \beta z_0}{\alpha r_0(1+\beta^2)},$$
$$d_1 = -\frac{\alpha\beta + \alpha^2 z_0 + z_0\beta^2(1+\alpha^2)}{\alpha r_0(1+\beta^2)}.$$

The comparison with macrospin simulations with $J_{\mathrm{app}} = 1 \times 10^{11}$ A/m^2 is shown in Fig. 5. The analyzer spin polarization η_{LONG} is $\eta_{\mathrm{LONG}} = r \times \eta_{\mathrm{PERP}} = 0.2r$, where r is the ratio p_x/p_z that varies from 0 to 80 in the simulations. This should give values of η_{LONG} larger than 1, which are unphysical, but they correspond to physical situations by rescaling the applied current density. Hence the absolute value of η_{LONG} is not so important, and only r matters. Figures 5(a) and 5(b) show m_z versus time for $r = 10$ ($\xi = 0.2$) and $r = 20$ ($\xi = 0.4$) respectively. For $r = 10$, the analytical expression is offset compared to the simulations, showing that there must be a
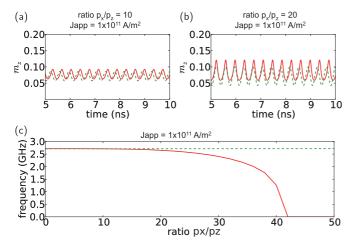

FIG. 5. (Color online) Macrospin simulation (red full line) and analytical expression at first order in ξ (green dashed line). (a) and (b) Out-of plane component of the magnetization m_z vs time, with applied current density of $J_{\mathrm{app}} = 1 \times 10^{11}$ A/m^2 ($z_0 = 0.06$) and with a ratio $r = p_x/p_z$ (a) $r = 10$ ($\xi = 0.2$) and (b) $r = 20$ ($\xi = 0.4$). (c) Frequency of the in-plane component m_x vs ratio p_x/p_z.

constant term arising from the second-order development that is not taken into account here. For $r = 20$, the simulated m_z is not sinusoidal, giving rise to a frequency mismatch with the first-order analytical expression. The frequency change versus ratio r shown in Fig. 5(c) shows that it is at least a third-order perturbation, as second-order frequency change cannot explain the frequency drop for large r.

Because the perturbation in first order has a frequency that is the fundamental frequency, like for the applied field, the second harmonic amplitude is nonzero:

$$m_x \approx r_0 \cos\tau + 2\xi \cos(2\tau + \psi').$$

Here, $\beta \ll 1$ and $z_0^2 \ll 1$ were assumed, and ψ' is a constant phase.

D. Interplay of two contributions

The power series expansion was only carried up to second order, because of the difficulty of developing further the analytical solutions and also because second order is sufficient regarding the negligible size of other terms. So far, only one contribution was considered at a time, but if we consider two simultaneous contributions, the perturbed expressions are changed. In first order, the influence of the different contributions can be added directly to determine the analytical expression for the angle θ and ϕ, there is no interaction between the contributions. However, at second order, the different contributions do not simply add together; there is also an additional term of interaction. Sometimes the interaction term vanishes, for instance, with the uniaxial anisotropy combined with the longitudinal spin torque or the applied field. However, if the longitudinal spin torque p_x and the applied field H_x (similar for H_y) are taken into account, an interaction term is added to the perturbed frequency.

We perform an expansion by setting $\varsigma = \frac{H_x}{p_x}$ and expanding with respect to ξ. The frequency correction is given by the term coming from H_x alone multiplied by v_x^2, the term coming from p_x alone multiplied by ξ^2, plus the interaction term, multiplied by ξv_x:

$$f_2 = -\frac{(1+\alpha^2)^3 z_0^2}{\alpha(1+\alpha^2 z_0^2)(\alpha^2 + z_0^2)}.$$

We can also treat the case of an angle-dependant spin torque amplitude with uniaxial anisotropy. First, we replace p_z by $p_z/(1 + \lambda_z \cos\theta)$ and treat λ_z as a perturbative parameter. Without any other perturbation, the variation of the frequency was derived exactly in the high-symmetry case, $f = f_0/(1 + \lambda_z z_+)$, and its Taylor expansion versus λ_z corresponds to the terms derived by the technique used here. However, there is an interaction with the uniaxial anisotropy term at the second order. On top of the uniaxial anisotropy and the spin torque asymmetry, $2\lambda_z^2 z_0^2$, terms, there is an interaction term multiplied by $\lambda_z \epsilon$:

$$f_2 = z_0^2(1 + \alpha^2).$$

Finally, we study the case of an angle-dependent spin torque along the x axis. There is an interaction between the angle-independent term p_x and the asymmetry parameter λ_x. For this we use the Taylor expansion versus λ_x of the full expression $\frac{p_x \sin\theta \cos\phi}{1 + \lambda_x \sin\theta \cos\phi}$. We compute the perturbation approximation by

expanding with the parameter ξ, and by λ_x if necessary. The frequency change is given by the term on ξ only, plus a first-order and a second-order terms depending on λ_x also:

$$f \approx f_0 \left[1 + \xi \lambda_x \frac{z_0}{2\alpha}(1 + \alpha^2) + \xi^2 \lambda_x^2 \frac{z_0^2}{4\alpha^2}(1 + \alpha^2)^2 \right.$$
$$\left. + \xi^2 \frac{(1 + \alpha^2)^2 z_0^2}{2(1 + \alpha^2 z_0^2)} \right].$$

IV. CRITICAL CURRENTS

A. Stability criterion

The expressions of the OPP trajectory with perturbation up to first order define a stability criterion for these trajectories. Regarding the Poincaré's map multiplier for the trajectories up to first order, $|\mu_1| < 1$ for all the perturbations treated in this paper, regardless of the amplitude of the perturbation. It means that the corrected trajectories are stable to sufficiently small thermal fluctuations. However, the multiplier does not give any information on the maximum amplitude of the perturbations up to which the out-of-plane precession is maintained.

For this we use the Melnikov function.[8,21] This function can be extracted from the equation of motion (4). By operating the vectorial product of this equation by $d\boldsymbol{m}$, then taking the scalar product with $\boldsymbol{m}$, we get

$$(\boldsymbol{m} \times \nabla_{\boldsymbol{m}}\mathcal{G} + \nabla_{\boldsymbol{m}}\mathcal{H}) \cdot d\boldsymbol{m} = 0.$$

If the trajectory of the system is periodic, the closed integral of the Hamiltonian part of the previous expression vanishes, namely,

$$\oint_\Gamma \nabla_{\boldsymbol{m}}\mathcal{H} \cdot d\boldsymbol{m} = 0.$$

Only the second part of the expression remains, $\oint (\boldsymbol{m} \times \nabla_{\boldsymbol{m}}\mathcal{G}) \cdot d\boldsymbol{m} = 0$. Using the expression of $d\boldsymbol{m}$ from the equation of motion, we obtain

$$\mathcal{M}(\Gamma) = \oint_\Gamma [\nabla_{\boldsymbol{m}}\mathcal{H} \cdot \nabla_{\boldsymbol{m}}\mathcal{G} - (\boldsymbol{m} \cdot \nabla_{\boldsymbol{m}}\mathcal{H})(\boldsymbol{m} \cdot \nabla_{\boldsymbol{m}}\mathcal{G})] \, dt = 0.$$

The above integral is called the Melnikov function. If the path Γ follows a periodic orbit solution of the equation of motion, then the Melnikov function vanishes. In the framework of weakly perturbed time periodic Hamiltonian systems, the Melnikov function is approximated by evaluating the closed loop integral over the trajectory Γ_0 of the unperturbed system ($\mathcal{G} = 0$) with constant energy w. The condition that the Melnikov function must vanish for a periodic orbit gives a constraint on the parameters of the system, namely the current density J.

The issue of calculating the critical parameters for which the periodic orbit is maintained has already been addressed by Ebels *et al.*[8] in the case of a free layer with a perpendicular polarizer and an applied field along the easy axis (x axis). They use a similar technique of zeroing the Melnikov function to obtain the critical current J_{c4} below which no periodic orbit is possible. It is important to be precise that the present paper focuses on determining the critical current J_{c4} and not J_{c1}, of which the origin and expression come from stability conditions of the in-plane equilibrium state. Ebels *et al.*[8] extracted numerical data for the critical current J_{c4} with respect

to the applied field, the anisotropy field H_k (named H_u in Ref. 8), the saturation magnetization M_S, and the damping constant α. Unfortunately, no analytical expression was found.

Critical currents for stability of the periodic out-of-plane precession can be derived using the expressions computed in the previous section. Using the expressions of the angles θ and ϕ at first order in the perturbation parameter, we can check that the in-plane angle variation $\dot\phi$ never vanishes. Otherwise, the precessional motion would be stopped. Using this criterion, critical spin torque amplitudes were computed, with only one perturbation at a time. The critical current J_c is proportional to the critical spin torque amplitudes p_{z_c}. The critical spin torque amplitudes are given beneath for the different perturbation cases, respectively, uniaxial anisotropy, in-plane applied field (there is no distinction between H_x and H_y direction, so we use H_a pointing in any arbitrary in-plane direction), and spin torque from the longitudinal polarizer. Here, we considered the case where $\alpha^2 \ll (H_a/M_s, H_k/M_s)$, which is more common in experiments, so that $\beta \ll 1$.

For the uniaxial anisotropy and the applied field, the out-of-plane precessional state is allowed as long as $p_z > p_{z_c}$. For these two perturbations, the stability criterion defines a minimal spin torque amplitude for the existence of the precessional state:

$$\text{anisotropy:} \quad p_{z_c} = \frac{\alpha}{2}\sqrt{H_k M_S}, \quad (26)$$

$$\text{in-plane applied field:} \quad p_{z_c} = \alpha\sqrt{H_a M_S}. \quad (27)$$

Figures 6 and 7 present the average in-plane component m_x of the free-layer magnetization with an applied current pulse of 10 ns from macrospin simulations. The average is taken between 8 and 10 ns to avoid the transient regime. The magnetization is initially almost completely out-of-plane ($m_z \approx 1$) to favor the OPP steady state that appears when the magnetization relaxes towards the plane. The average magnetization is represented versus the applied current density, the anisotropy field in Fig. 6, and the in-plane applied field along $\boldsymbol{u}_x$ in Fig. 7. The blue and red colors correspond to an average magnetization in equilibrium in the plane, $\langle m_x \rangle = \pm 1$, respectively. The light green color corresponds

FIG. 6. (Color online) From macrospin simulations, the average in-plane component m_x of the magnetization in permanent regime vs applied current density and anisotropy field. Dark blue and red colors correspond to the equilibrium states $\langle m_x \rangle = \pm 1$, respectively. Light green color corresponds to $\langle m_x \rangle = 0$, the out-of-plane precessional steady state. The analytical critical current is represented by the black dashed line.
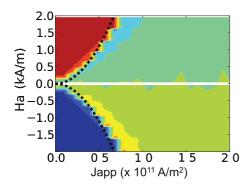
FIG. 7. (Color online) Same as Fig. 6 with the applied field along $\boldsymbol{u}_x$.

FIG. 9. (Color online) Same as Fig. 6 with the ratio $r = p_x/p_z$.

to $\langle m_x \rangle = 0$, representative of the OPP steady state. We can extract the critical current J_{c4} from the boundary between light green and dark blue/red regions. The analytical critical line from expression (27) is represented in black dashed line. For the highest value of anisotropy and applied fields used in the simulations, the analytical critical current differs from the simulations value by 20% and 10%, respectively. Notice that for the anisotropy figure, the equilibrium region is not uniformly colored and presents blue and red stripes because depending on where the OPP is stopped, the magnetization relaxes to $m_x = -1$ or 1. The final states actually depend on the transient regime, they are of little interest here. The critical currents obtained by finding numerically the root of the Melnikov function, as in Ebels *et al.*,[8] are also in agreement with the analytical expressions. In fact, they match with an error of less than 10% as seen on Fig. 8, which represents the critical spin torque amplitude find by numerical methods, normalized by the analytical expression, versus applied field.

For the analyzer spin torque, the same criterion used for anisotropy and applied field defines a maximal spin torque amplitude for which the periodic orbit exists. The critical current density depends on the ratio $r = p_x/p_z$ of the two spin torque contributions:

$$p_{zc} = \alpha M_S(1 - \alpha^2 r^2) \qquad \text{for} \quad 2\alpha r < 1,$$
$$p_{zc} = \frac{\alpha^2 M_S}{\sqrt{4\alpha^2 r^2 - 1}} \qquad \text{for} \quad 2\alpha r > 1. \tag{28}$$

The phase diagram of Fig. 9 represents the average of the in-plane magnetization m_x from macrospin simulations with different applied current density and ratio $r = px/pz$. The analytical expression for the critical current density is drawn in black dashed line. The analytical critical line corresponds
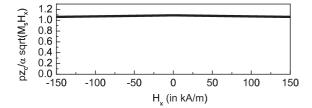

FIG. 8. Critical spin torque amplitude computed with the Melnikov function vs applied field H_x along $\boldsymbol{u}_x$. The critical spin torque amplitude is normalized by the analytical expression.
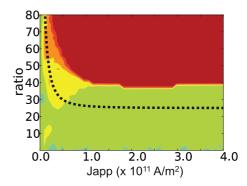
qualitatively to the region boundary of the diagram. However, for a given current density, the corresponding critical ratio is 35% smaller with the analytical expression than found by macrospin simulations. On the critical line, the ratio is around 40, giving a perturbation parameter $\xi = 0.8$. In this range, we should consider higher orders to obtain a better accuracy on the critical current.

V. DISCUSSION

The perturbative approach developed here shows good agreement with macrospin simulations even at low order. For the anisotropy and the applied fields, we were able to calculate the frequency change due to these perturbations, except close to the range where the frequency vanishes. For these perturbations, we were also able to give a good approximation of the critical current J_{c4} at which the OPP state disappears. In fact, the perturbation parameter, for instance ϵ, is more or less inversely proportional to the frequency of the OPP, so it is clear that the perturbative expansion is less precise when the frequency vanishes, or close to the critical current, which is determined by this same criterion. Another limit is when the applied current density tends towards zero, that corresponds also to a vanishing frequency. However, this limit is not very relevant physically as the dynamical system studied here presents another critical current density, usually called J_{c1}, below which the magnetization remains in static equilibrium in-plane. So the dynamical system presents a bifurcation between an equilibrium and a limit cycle not associated with this equilibrium, and the frequency is nonvanishing close to the critical current J_{c1}.

This study has been carried out for an in-plane magnetized MTJ with additional perpendicular polarizer. The same perturbation expansion could also be done with a MTJ with in-plane precession (IPP) steady state. A good starting point to treat this case would be to consider the Landau-Lifshitz-Gilbert equation in complex numbers as described by the Kim, Tiberkevich, and Slavin (KTS) theory;[10–12] if we write the complex differential equation as a couple of differential equations for the amplitude and the phase, the amplitude would play the role of θ here, and the phase the role of ϕ. By rescaling the time with the eigenfrequency ω_0 corresponding to the FMR frequency, we expect to carry out the Lindstedt's series expansion of the trajectory with constant energy, considering α as a small parameter. It should give a good approximation of the nonlinear parameters modifying the

frequency and the linewidth, and also the relative amplitude of the second harmonics, which are not treated by the present theory. However, this is beyond the scope of this paper.

APPENDIX: LINDSTEDT'S SERIES EXPANSION

The Lindstedt series expansion is a method used to obtain an approximation of the solution of a dynamical system (S) with a small perturbation, by using the solution of an unperturbed system (S_0). Let (S) and (S_0) be the following systems, and $\boldsymbol{X}_0$ be the solution of (S_0):

$$(S): \quad \dot{\boldsymbol{X}} = \boldsymbol{F}(\boldsymbol{X}) + \epsilon \boldsymbol{G}(\boldsymbol{X}),$$
$$(S_0): \quad \dot{\boldsymbol{X}_0} = \boldsymbol{F}(\boldsymbol{X}_0).$$

In order to find the solution of (S), we suppose that the solution $\boldsymbol{X}$ of (S) can be written as a power series of the small parameter ϵ:

$$\boldsymbol{X} = \boldsymbol{X}_0 + \epsilon \boldsymbol{X}_1 + \epsilon^2 \boldsymbol{X}_2 + \dots .$$

We can then replace $\boldsymbol{X}$ by the following power series in the system (S):

$$\dot{\boldsymbol{X}}_0 + \epsilon \dot{\boldsymbol{X}}_1 + \epsilon^2 \dot{\boldsymbol{X}}_2 + \dots = \boldsymbol{F}(\boldsymbol{X}_0 + \epsilon \boldsymbol{X}_1 + \epsilon^2 \boldsymbol{X}_2 + \dots)$$
$$+ \epsilon \boldsymbol{G}(\boldsymbol{X}_0 + \epsilon \boldsymbol{X}_1 + \epsilon^2 \boldsymbol{X}_2 + \dots).$$

Then the perturbed dynamical system (S) becomes (S_1) at the first order in ϵ:

$$(S_1): \quad \dot{\boldsymbol{X}}_0 + \epsilon \dot{\boldsymbol{X}}_1 = \boldsymbol{F}(\boldsymbol{X}_0) + \epsilon \, \mathrm{d}\boldsymbol{F}(\boldsymbol{X}_0)[\boldsymbol{X}_1] + \epsilon \boldsymbol{G}(\boldsymbol{X}_0).$$

Here, $\mathrm{d}\boldsymbol{F}(\boldsymbol{X}_0)$ is the differential of the vector field $\boldsymbol{F}$ evaluated at $\boldsymbol{X}_0$. By identification, we obtain two differential equations, where the first one is the unperturbed system (S_0):

$$\dot{\boldsymbol{X}}_0 = \boldsymbol{F}(\boldsymbol{X}_0),$$
$$\dot{\boldsymbol{X}}_1 = \mathrm{d}\boldsymbol{F}(\boldsymbol{X}_0)[\boldsymbol{X}_1] + \boldsymbol{G}(\boldsymbol{X}_0).$$

This method theoretically allows to calculate the solution up to any order. However, if the solution $\dot{\boldsymbol{X}}_k$ at order k is periodic, the resolution of the dynamical system at order $k+1$ may give rise to nonperiodic secular terms, often $t \cos(\omega_0 t)$ terms, or, for the equation addressed in this paper, t terms. In order to get rid of the secular terms, we also suppose that the angular frequency can be written as a power series of ϵ: $\omega = \omega_0 + \epsilon \omega_1 + \epsilon^2 \omega_2 + \dots$. By rescaling time with the change of variables $\tau = \omega t$, we rewrite system (S) as

$$(\omega_0 + \epsilon \omega_1 + \epsilon^2 \omega_2 + \dots)(\dot{\boldsymbol{X}}_0 + \epsilon \dot{\boldsymbol{X}}_1 + \epsilon^2 \dot{\boldsymbol{X}}_2 + \dots)$$
$$= \boldsymbol{F}(\boldsymbol{X}_0 + \epsilon \boldsymbol{X}_1 + \epsilon^2 \boldsymbol{X}_2 + \dots)$$
$$+ \epsilon \boldsymbol{G}(\boldsymbol{X}_0 + \epsilon \boldsymbol{X}_1 + \epsilon^2 \boldsymbol{X}_2 + \dots). \tag{A1}$$

Henceforth, $\dot{\boldsymbol{X}}$ denotes a derivation with respect to τ. The first-order system becomes

$$\dot{\boldsymbol{X}}_1 + \omega_1 \dot{\boldsymbol{X}}_0 = \mathrm{d}\boldsymbol{F}(\boldsymbol{X}_0)[\boldsymbol{X}_1] + \boldsymbol{G}(\boldsymbol{X}_0).$$

The value of ω_1, and higher-order terms ω_k, is set so that secular terms are canceled. In the majority of the cases treated in this paper, it appears that $\omega_1 = 0$ because no secular terms appear at order 1. However, at order 2, in order to cancel the secular terms, ω_2 needs to be nonzero.

However, this way of solving the perturbed system does not account for an additional temporal dephasing between the angles θ and ϕ. According to the exact solutions of Eq. (8), the second term of the right-hand side from the expression of ϕ is a dephasing term that accounts for the initial condition z_i. In this particular case, the angle θ is constant in the permanent regime so this dephasing has no effect. However, solving a perturbed system gives rise to oscillating terms on the angle θ, for which the exact dephasing with the angle ϕ is important, specially for second-order terms. In fact, the second-order terms partly account for an interaction between θ and ϕ harmonics. In order to solve this issue, we first calculate the equation of the angle θ with respect to ϕ, then the expression of $\theta(\tau)$ and $\phi(\tau)$. By dividing the first equation of Eq. (16) by the second equation, we obtain the differential equation governing $\theta(\phi)$:

$$\frac{\mathrm{d}\theta}{\mathrm{d}\phi} = \sin\theta \frac{g_\theta(\theta) + \epsilon f_\theta^\epsilon(\theta, \phi)}{h_\phi(\theta) + \epsilon f_\phi^\epsilon(\theta, \phi)}. \tag{A2}$$

Both sides of this equation can be Taylor expanded with respect to ϵ by using the expressions for θ and ϕ from Eq. (17). Hence the expression of θ with respect to ϕ is computed up to any given order k. The assumption that $\epsilon \ll 1$, or that the perturbation is small compared to the nonperturbated frequency, is critical here.

Thereafter, the expression of $\theta(\phi)$ at order k is injected in the differential equations (16) of θ and ϕ versus time τ to obtain the precession harmonics and the frequency changes, using the Lindstedt's series.

The results of Sec. III were obtained with the computer algebra system MAXIMA,[22] by differentiating Eq. (A1) with respect to ϵ to get the equations at the different orders in ϵ. At order 2, the solutions already hold many terms that are not shown here because we were only concerned about the frequency change. The computation of higher-order terms is rather complicated.

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